

Physics 4B

Solutions to Chapter 31 HW

Chapter 31: Questions: 2, 8, 12

Exercises & Problems: 2, 23, 24, 32, 41, 44, 48, 60, 72, 83

Question 31-2

(a) less; (b) greater

Question 31-8

(a) 1 and 4; (b) 2 and 3

Question 31-12

(a) lead; (b) capacitive; (c) less

Problem 31-2

(a) We recall the fact that the period is the reciprocal of the frequency. It is helpful to refer also to Fig. 31-1. The values of t when plate A will again have maximum positive charge are multiples of the period:

$$t_A = nT = \frac{n}{f} = \frac{n}{2.00 \times 10^3 \text{ Hz}} = n(5.00 \mu\text{s}),$$

where $n = 1, 2, 3, 4, \dots$. The earliest time is ($n = 1$) $t_A = 5.00 \mu\text{s}$.

(b) We note that it takes $t = \frac{1}{2}T$ for the charge on the other plate to reach its maximum positive value for the first time (compare steps a and e in Fig. 31-1). This is when plate A acquires its most negative charge. From that time onward, this situation will repeat once every period. Consequently,

$$t = \frac{1}{2}T + (n-1)T = \frac{1}{2}(2n-1)T = \frac{(2n-1)}{2f} = \frac{(2n-1)}{2(2 \times 10^3 \text{ Hz})} = (2n-1)(2.50 \mu\text{s}),$$

where $n = 1, 2, 3, 4, \dots$. The earliest time is ($n = 1$) $t = 2.50 \mu\text{s}$.

(c) At $t = \frac{1}{4}T$, the current and the magnetic field in the inductor reach maximum values for the first time (compare steps a and c in Fig. 31-1). Later this will repeat every half-period (compare steps c and g in Fig. 31-1). Therefore,

$$t_L = \frac{T}{4} + \frac{(n-1)T}{2} = (2n-1)\frac{T}{4} = (2n-1)(1.25 \mu\text{s}),$$

where $n = 1, 2, 3, 4, \dots$. The earliest time is ($n = 1$) $t = 1.25 \mu\text{s}$.

Problem 31-23

(a) The total energy U is the sum of the energies in the inductor and capacitor:

$$U = U_E + U_B = \frac{q^2}{2C} + \frac{i^2 L}{2} = \frac{(3.80 \times 10^{-6} \text{ C})^2}{2(7.80 \times 10^{-6} \text{ F})} + \frac{(9.20 \times 10^{-3} \text{ A})^2 (25.0 \times 10^{-3} \text{ H})}{2} = 1.98 \times 10^{-6} \text{ J}.$$

(b) We solve $U = Q^2/2C$ for the maximum charge:

$$Q = \sqrt{2CU} = \sqrt{2(7.80 \times 10^{-6} \text{ F})(1.98 \times 10^{-6} \text{ J})} = 5.56 \times 10^{-6} \text{ C}.$$

(c) From $U = I^2 L/2$, we find the maximum current:

$$I = \sqrt{\frac{2U}{L}} = \sqrt{\frac{2(1.98 \times 10^{-6} \text{ J})}{25.0 \times 10^{-3} \text{ H}}} = 1.26 \times 10^{-2} \text{ A}.$$

(d) If q_0 is the charge on the capacitor at time $t = 0$, then $q_0 = Q \cos \phi$ and

$$\phi = \cos^{-1}\left(\frac{q}{Q}\right) = \cos^{-1}\left(\frac{3.80 \times 10^{-6} \text{ C}}{5.56 \times 10^{-6} \text{ C}}\right) = \pm 46.9^\circ.$$

For $\phi = +46.9^\circ$ the charge on the capacitor is decreasing, for $\phi = -46.9^\circ$ it is increasing. To check this, we calculate the derivative of q with respect to time, evaluated for $t = 0$. We obtain $-\omega Q \sin \phi$, which we wish to be positive. Since $\sin(+46.9^\circ)$ is positive and $\sin(-46.9^\circ)$ is negative, the correct value for increasing charge is $\phi = -46.9^\circ$.

(e) Now we want the derivative to be negative and $\sin \phi$ to be positive. Thus, we take $\phi = +46.9^\circ$.

Problem 31-24

The charge q after N cycles is obtained by substituting $t = NT = 2\pi N/\omega'$ into Eq. 31-25:

$$\begin{aligned} q &= Qe^{-Rt/2L} \cos(\omega't + \phi) = Qe^{-RNT/2L} \cos[\omega'(2\pi N/\omega') + \phi] \\ &= Qe^{-RN(2\pi\sqrt{L/C})/2L} \cos(2\pi N + \phi) \\ &= Qe^{-N\pi R\sqrt{C/L}} \cos \phi. \end{aligned}$$

We note that the initial charge (setting $N = 0$ in the above expression) is $q_0 = Q \cos \phi$, where $q_0 = 6.2 \mu\text{C}$ is given (with 3 significant figures understood). Consequently, we write the above result as $q_N = q_0 \exp(-N\pi R\sqrt{C/L})$.

(a) For $N = 5$, $q_5 = (6.2 \mu\text{C}) \exp(-5\pi(7.2\Omega)\sqrt{0.0000032\text{F}/12\text{H}}) = 5.85 \mu\text{C}$.

(b) For $N = 10$, $q_{10} = (6.2 \mu\text{C}) \exp(-10\pi(7.2\Omega)\sqrt{0.0000032\text{F}/12\text{H}}) = 5.52 \mu\text{C}$.

(c) For $N = 100$, $q_{100} = (6.2 \mu\text{C}) \exp(-100\pi(7.2\Omega)\sqrt{0.0000032\text{F}/12\text{H}}) = 1.93 \mu\text{C}$.

Problem 31-32

(a) The circuit consists of one generator across one inductor; therefore, $\varepsilon_m = V_L$. The current amplitude is

$$I = \frac{\varepsilon_m}{X_L} = \frac{\varepsilon_m}{\omega_d L} = \frac{25.0 \text{ V}}{(377 \text{ rad/s})(12.7 \text{ H})} = 5.22 \times 10^{-3} \text{ A} .$$

(b) When the current is at a maximum, its derivative is zero. Thus, Eq. 30-35 gives $\varepsilon_L = 0$ at that instant. Stated another way, since $\varepsilon(t)$ and $i(t)$ have a 90° phase difference, then $\varepsilon(t)$ must be zero when $i(t) = I$. The fact that $\phi = 90^\circ = \pi/2$ rad is used in part (c).

(c) Consider Eq. 31-28 with $\varepsilon = -\varepsilon_m / 2$. In order to satisfy this equation, we require $\sin(\omega_d t) = -1/2$. Now we note that the problem states that ε is increasing *in magnitude*, which (since it is already negative) means that it is becoming more negative. Thus, differentiating Eq. 31-28 with respect to time (and demanding the result be negative) we must also require $\cos(\omega_d t) < 0$. These conditions imply that ωt must equal $(2n\pi - 5\pi/6)$ [$n = \text{integer}$]. Consequently, Eq. 31-29 yields (for all values of n)

$$i = I \sin\left(2n\pi - \frac{5\pi}{6} - \frac{\pi}{2}\right) = (5.22 \times 10^{-3} \text{ A}) \left(\frac{\sqrt{3}}{2}\right) = 4.51 \times 10^{-3} \text{ A} .$$

Problem 31-41

(a) The capacitive reactance is

$$X_C = \frac{1}{\omega_d C} = \frac{1}{2\pi f_d C} = \frac{1}{2\pi(60.0 \text{ Hz})(70.0 \times 10^{-6} \text{ F})} = 37.9 \Omega .$$

The inductive reactance 86.7Ω is unchanged. The new impedance is

$$Z = \sqrt{R^2 + (X_L - X_C)^2} = \sqrt{(200\Omega)^2 + (37.9\Omega - 86.7\Omega)^2} = 206\Omega .$$

(b) The phase angle is

$$\phi = \tan^{-1} \left(\frac{X_L - X_C}{R} \right) = \tan^{-1} \left(\frac{86.7 \Omega - 37.9 \Omega}{200 \Omega} \right) = 13.7^\circ.$$

(c) The current amplitude is

$$I = \frac{\varepsilon_m}{Z} = \frac{36.0 \text{ V}}{206 \Omega} = 0.175 \text{ A}.$$

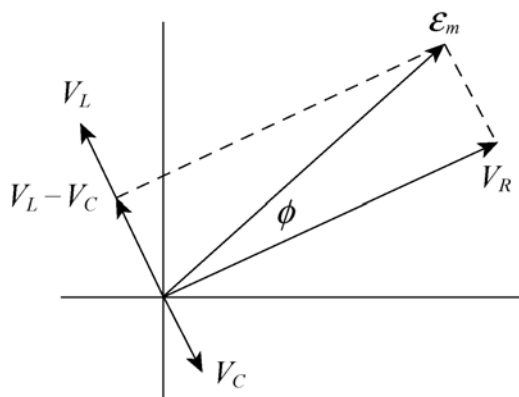
(d) We first find the voltage amplitudes across the circuit elements:

$$V_R = IR = (0.175 \text{ A})(200 \Omega) = 35.0 \text{ V}$$

$$V_L = IX_L = (0.175 \text{ A})(86.7 \Omega) = 15.2 \text{ V}$$

$$V_C = IX_C = (0.175 \text{ A})(37.9 \Omega) = 6.62 \text{ V}$$

Note that $X_L > X_C$, so that ε_m leads I . The phasor diagram is drawn to scale below.



Problem 31-44

(a) The capacitive reactance is

$$X_C = \frac{1}{2\pi fC} = \frac{1}{2\pi(400 \text{ Hz})(24.0 \times 10^{-6} \text{ F})} = 16.6 \Omega.$$

(b) The impedance is

$$\begin{aligned} Z &= \sqrt{R^2 + (X_L - X_C)^2} = \sqrt{R^2 + (2\pi fL - X_C)^2} \\ &= \sqrt{(220 \Omega)^2 + [2\pi(400 \text{ Hz})(150 \times 10^{-3} \text{ H}) - 16.6 \Omega]^2} = 422 \Omega. \end{aligned}$$

(c) The current amplitude is

$$I = \frac{\varepsilon_m}{Z} = \frac{220 \text{ V}}{422 \Omega} = 0.521 \text{ A}.$$

(d) Now $X_C \propto C_{\text{eq}}^{-1}$. Thus, X_C increases as C_{eq} decreases.

(e) Now $C_{\text{eq}} = C/2$, and the new impedance is

$$Z = \sqrt{(220 \Omega)^2 + [2\pi(400 \text{ Hz})(150 \times 10^{-3} \text{ H}) - 2(16.6 \Omega)]^2} = 408 \Omega < 422 \Omega .$$

Therefore, the impedance decreases.

(f) Since $I \propto Z^{-1}$, it increases.

Problem 31-48

(a) With both switches closed (which effectively removes the resistor from the circuit), the impedance is just equal to the (net) reactance and is equal to

$$X_{\text{net}} = (12 \text{ V})/(0.447 \text{ A}) = 26.85 \Omega .$$

With switch 1 closed but switch 2 open, we have the same (net) reactance as just discussed, but now the resistor is part of the circuit; using Eq. 31-65 we find

$$R = \frac{X_{\text{net}}}{\tan \phi} = \frac{26.85 \Omega}{\tan 15^\circ} = 100 \Omega .$$

(b) For the first situation described in the problem (both switches open) we can reverse our reasoning of part (a) and find

$$X_{\text{net first}} = R \tan \phi' = (100 \Omega) \tan(-30.9^\circ) = -59.96 \Omega .$$

We observe that the effect of switch 1 implies

$$X_C = X_{\text{net}} - X_{\text{net first}} = 26.85 \Omega - (-59.96 \Omega) = 86.81 \Omega .$$

Then Eq. 31-39 leads to $C = 1/\omega X_C = 30.6 \mu\text{F}$.

(c) Since $X_{\text{net}} = X_L - X_C$, then we find $L = X_L/\omega = 301 \text{ mH}$.

Problem 31-60

The current in the circuit satisfies $i(t) = I \sin(\omega_d t - \phi)$, where

$$\begin{aligned}
 I &= \frac{\mathcal{E}_m}{Z} = \frac{\mathcal{E}_m}{\sqrt{R^2 + (\omega_d L - 1/\omega_d C)^2}} \\
 &= \frac{45.0 \text{ V}}{\sqrt{(16.0 \Omega)^2 + \left\{ (3000 \text{ rad/s})(9.20 \text{ mH}) - 1/\left[(3000 \text{ rad/s})(31.2 \mu\text{F}) \right] \right\}^2}} \\
 &= 1.93 \text{ A}
 \end{aligned}$$

and

$$\begin{aligned}
 \phi &= \tan^{-1} \left(\frac{X_L - X_C}{R} \right) = \tan^{-1} \left(\frac{\omega_d L - 1/\omega_d C}{R} \right) \\
 &= \tan^{-1} \left[\frac{(3000 \text{ rad/s})(9.20 \text{ mH})}{16.0 \Omega} - \frac{1}{(3000 \text{ rad/s})(16.0 \Omega)(31.2 \mu\text{F})} \right] \\
 &= 46.5^\circ.
 \end{aligned}$$

(a) The power supplied by the generator is

$$\begin{aligned}
 P_g &= i(t)\mathcal{E}(t) = I \sin(\omega_d t - \phi) \mathcal{E}_m \sin \omega_d t \\
 &= (1.93 \text{ A})(45.0 \text{ V}) \sin \left[(3000 \text{ rad/s})(0.442 \text{ ms}) \right] \sin \left[(3000 \text{ rad/s})(0.442 \text{ ms}) - 46.5^\circ \right] \\
 &= 41.4 \text{ W}.
 \end{aligned}$$

(b) With

$$v_c(t) = V_c \sin(\omega_d t - \phi - \pi/2) = -V_c \cos(\omega_d t - \phi)$$

where $V_c = I/\omega_d C$, the rate at which the energy in the capacitor changes is

$$\begin{aligned}
 P_c &= \frac{d}{dt} \left(\frac{q^2}{2C} \right) = i \frac{q}{C} = i v_c \\
 &= -I \sin(\omega_d t - \phi) \left(\frac{I}{\omega_d C} \right) \cos(\omega_d t - \phi) = -\frac{I^2}{2\omega_d C} \sin \left[2(\omega_d t - \phi) \right] \\
 &= -\frac{(1.93 \text{ A})^2}{2(3000 \text{ rad/s})(31.2 \times 10^{-6} \text{ F})} \sin \left[2(3000 \text{ rad/s})(0.442 \text{ ms}) - 2(46.5^\circ) \right] \\
 &= -17.0 \text{ W}.
 \end{aligned}$$

(c) The rate at which the energy in the inductor changes is

$$\begin{aligned}
 P_L &= \frac{d}{dt} \left(\frac{1}{2} Li^2 \right) = Li \frac{di}{dt} = LI \sin(\omega_d t - \phi) \frac{d}{dt} [I \sin(\omega_d t - \phi)] = \frac{1}{2} \omega_d LI^2 \sin[2(\omega_d t - \phi)] \\
 &= \frac{1}{2} (3000 \text{ rad/s}) (1.93 \text{ A})^2 (9.20 \text{ mH}) \sin[2(3000 \text{ rad/s})(0.442 \text{ ms}) - 2(46.5^\circ)] \\
 &= 44.1 \text{ W}.
 \end{aligned}$$

(d) The rate at which energy is being dissipated by the resistor is

$$\begin{aligned}
 P_R &= i^2 R = I^2 R \sin^2(\omega_d t - \phi) = (1.93 \text{ A})^2 (16.0 \Omega) \sin^2[(3000 \text{ rad/s})(0.442 \text{ ms}) - 46.5^\circ] \\
 &= 14.4 \text{ W}.
 \end{aligned}$$

(e) Equal. $P_L + P_R + P_C = 44.1 \text{ W} - 17.0 \text{ W} + 14.4 \text{ W} = 41.5 \text{ W} = P_g$.

Problem 31-72

(a) From Eq. 31-65, we have

$$\phi = \tan^{-1} \left(\frac{V_L - V_C}{V_R} \right) = \tan^{-1} \left(\frac{V_L - (V_L / 1.50)}{(V_L / 2.00)} \right)$$

which becomes $\tan^{-1}(2/3) = 33.7^\circ$ or 0.588 rad .

(b) Since $\phi > 0$, it is inductive ($X_L > X_C$).

(c) We have $V_R = IR = 9.98 \text{ V}$, so that $V_L = 2.00V_R = 20.0 \text{ V}$ and $V_C = V_L/1.50 = 13.3 \text{ V}$. Therefore, from Eq. 31-60, we have

$$\varepsilon_m = \sqrt{V_R^2 + (V_L - V_C)^2} = \sqrt{(9.98 \text{ V})^2 + (20.0 \text{ V} - 13.3 \text{ V})^2} = 12.0 \text{ V}.$$

Problem 31-83

From Eq. 31-4 we get $f = 1/2\pi\sqrt{LC} = 1.84 \text{ kHz}$.